THEORETICAL STUDY ON SLOW-LIGHT GENERATED BY INTEGRATED MICRORING RESONATOR WITH WIDE BANDWIDTH AND HIGH GAIN

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Graphical abstract

Abstract

We propose a new approach to generate slow light transmission with large bandwidth and high buildup factor by using a soliton pulse propagating within integrated ring resonator circuit. The system consisted series of micron-size ring resonator fabricated by using nonlinear InGaAsP/InP material that are laterally coupled together. For convenience of analysis, optical transfer function for this model is obtained by using z-transform method. Slow light performances were modeled and discuss in this paper. Intensity buildup induced within the series of rings located at left and right sides of the system while strong nonlinear Kerr effect and mutual coupling leads to the spreading frequency bands within the device. Numerical simulation verifies that signal pulse with 45 ps relative delay time and bandwidth of 5.9 GHz (47 pm) are obtained at the communication wavelength around 1550 nm for a 100 ps signal pulse.

Keywords: Nonlinear optics, fibers, coupled resonators, slow-light

Abstrak

Kami mencadangkan pendekatan baru untuk menjana cahaya perlahan dengan jalur lebar yang besar dan faktor tokokan tinggi dengan menggunakan denyut soliton merambat dalam litar bersepadu cincin resonator. Sistem ini terdiri dari cincin resonator bersaiz mikron direka dengan menggunakan bahan tak linear InGaAsP / InP yang dicantumkan secara sisipan. Untuk kemudahan analisis, fungsi pemindahan optik untuk model ini diperolehi dengan menggunakan kaedah z-transform. Prestasi cahaya perlahan dimodelkan dan dibincangkan dalam manuskrip ini. Intensiti tokokan terhasil dalam siri cincin yang terletak di kiri dan kanan sistem manakala kesan tak linear Kerr dan gandingan bersama membawa kepada rebakan jalur frekuensi dalam peranti. Simulasi berangka mengesahkan bahawa nadi isyarat dengan 45 ps masa tunda relatif dan lebar jalur 5.9 GHz (47 pm) diperolehi pada panjang gelombang komunikasi sekitar 1.550 nm untuk isyarat denyut 100 ps.

Kata kunci: Optik tak-linear, fiber, pengalun bercantum, cahaya perlahan
1.0 INTRODUCTION

In recent years, slow light technologies have become one of the most important research topics with much of the studies focusing on development of techniques that has the ability to control the transmission delay of optical signals [1, 2]. This technique offers wide range of potential applications in signal processing such as optical buffer [3], optical delay lines [4, 5] and quantum information [6]. Among all the approaches in realization of slow light, integrated optical waveguide platform is more promising due to its strong optical confinement and easy integration with ultra-compact photonics circuit [7, 8]. Recent studies have shown that strong mutual coupling in integrated multi-coupled ring resonator system induced resonant splitting that leads to pulse advancement or delay with relatively low distortion in signal pulse [9]. For instance, a number of theoretical and experimental works on generation of slow light have been demonstrated [10, 11]. Technically, the use of ring resonator device requires a high degree of control especially in all-optical signal processing system. However, the fundamental of single ring resonator system only provide simple Lorentzian response [12]. Due to this matter, some configuration is not suitable for many important applications in dense wavelength division multiplexing (DWDM) systems. Therefore, tailoring the filter response shape is needed in order to enhance the potential application of ring resonator in various field of studies. For this purpose, specific design of microring resonator (MRR) device consisting one center ring coupled to a series of microrings on the left and right hand side of the system is considered as shown in Figure 1. This unique configuration is known as “double-PANDA” ring resonator device. Optical transfer function of such system is first derived by implementing z-transform method to the photonics circuit accordingly. To date, different approaches have been used in order to determine the signal transfer function of integrated MRR system including transfer matrix method [13-17], Vernier principle [18-21] and signal flow graph method [22, 23]. In this study, a novel attempt has been made in order to achieve a delay light transmission within the proposed system with relatively large bandwidth and high buildup factor. The whole research works might contribute towards the new perspective on realization of slow light.

2.0 OPERATING PRINCIPLE

Operating system of double PANDA ring resonator is shown in Figure 1. The system consisted of one center ring, two straight waveguides and four side-rings. The straight waveguides are coupled on top and bottom of the center ring, while all side-rings are arranged in particular configuration such that two rings are serially coupled on both lateral sides of the center ring by using 2x2 optical directional coupler. This configuration consisted of four main ports. Optical signals are injected into the system through input and add ports while the corresponding signals are ejected from the system through throughput and drop ports respectively.

![Figure 1 (a) Double PANDA ring resonator configuration and (b) layer structure of composed system](image)

3.0 PHOTONICS TRANSFER FUNCTION

Referring to the schematic diagram of double PANDA ring resonator system as depicted in Figure 2, input signals $E_{in}$ and $E_{add}$ are injected into the system through input and add ports. Meanwhile, output signals $E_1$ and $E_4$ are ejected from the system via throughput and drop ports. Circulating fields associated with different points on the center ring are denoted as $E_0$, $E_i$, $E_s$, $E_x$, $E_5$, $E_6$ and $E_T$. $\kappa$ represents the coupling factor for the optical directional coupler connecting ring resonator and the bus waveguide. Fraction of light that continues to propagate in its original direction is called the transmission coefficient and it is given by $\gamma_i = \sqrt{|1-\kappa|}$. Vice versa, the other fraction which travels across the waveguide (coupled to the second waveguide) is called cross-coupling coefficient denoted by $j\sqrt{\kappa}$ [24, 25]. Transmission for one complete roundtrip of light propagation in ring resonator is derived using z-transform method and given it is by $\xi=az^{n-1}$ [26]. The term $a$ is assigned as the total loss coefficient of the ring waveguide for one complete roundtrip and it is given by $a=\exp(-\alpha L)$, where $\alpha$ refers to attenuation loss constant. Noted that if we assume that ring resonator is ideal with no losses, then the value of $a$ is unity. z-transform parameter is represented by $(1-az^{-1})^{-1}$ and propagation constant and $L=2\pi R$ is circumference of the ring. The term $z-1$ refers...
to the phase propagation of field passing through the ring resonator system [27].

As shown in Figure 2, \( \kappa_1 \) and \( \kappa_2 \) are the coupler parameters for the coupling between center ring and the straight waveguides while \( \kappa_3, \kappa_4, \kappa_5 \) and \( \kappa_6 \) are the coupler parameters for the coupling between the rings respectively. Radius of the center ring is denoted as \( R \) while \( R_1, R_2, R_3 \) and \( R_4 \) are the ring radiiuses for the side-rings on the left and right sides of the center ring. To obtain the optical transfer function of this system, new set of parameters are used for simplification purpose which are \( y_1 = \sqrt{1 - \kappa_1}, y_2 = \sqrt{1 - \kappa_2}, y_3 = \sqrt{1 - \kappa_3}, y_4 = \sqrt{1 - \kappa_4}, y_5 = \sqrt{1 - \kappa_5}, y_6 = \sqrt{1 - \kappa_6} \) and \( x_1 = \sqrt{1 - y_1}, x_2 = \sqrt{1 - y_2}, x_3 = \sqrt{1 - y_3}, x_4 = \sqrt{1 - y_4}, x_5 = \sqrt{1 - y_5}, x_6 = \sqrt{1 - y_6} \) where \( y_i \) is the \( i \)th intensity insertion loss coefficients for six different coupler of the system. By taking into account the coupling coefficient and insertion loss coefficient for each coupler, throughput and drop port signals can be expressed by:

\[
E_t = x_1 y_1 E_{in} + j x_1 \sqrt{\kappa_3} E_1
\]

\[
E_d = x_2 y_2 E_{add} + j x_2 \sqrt{\kappa_3} E_3
\]

For further simplification, new set of parameters are assigned as \( \varphi = \alpha L / \sqrt{k_n L} \) where \( \varphi_{R1} = - (\alpha L _R1) / 2 \sqrt{k_n L_R1} \), \( \varphi_{R2} = - (\alpha L _R2) / 2 \sqrt{k_n L_R2} \), \( \varphi_{R3} = - (\alpha L _R3) / 2 \sqrt{k_n L_R3} \) and \( \varphi_{R4} = - (\alpha L _R4) / 2 \sqrt{k_n L_R4} \) which indicate the transmission of light after passing through one complete roundtrip for each corresponding ring. To obtain the analytical transfer function of this system, let first consider the right-hand side part of the circuit containing \( R_1 \) and \( R_2 \). By using \( z \)-transform method, circulating fields on the right rings as shown in schematic diagram in Figure 3 are expressed by:

\[
E_a = j x_3 \sqrt{\kappa_3} E_1 + x_4 y_3 E_f
\]

\[
E_b = E_a \exp \left( \frac{\varphi_{R1}}{2} \right)
\]

\[
E_c = j x_5 \sqrt{\kappa_5} E_b + x_6 y_5 E_d
\]

\[
E_d = E_c \exp (\varphi_{R2})
\]

\[
E_e = j x_5 \sqrt{\kappa_5} E_d + x_6 y_5 E_b
\]

\[
E_f = E_e \exp \left( \frac{\varphi_{R1}}{2} \right)
\]

Simplifying these equations yield:

\[
E_a = AE_1
\]

Where

\[
A = \frac{j x_3 \sqrt{\kappa_3}}{1 - x_3 y_3 \frac{x_5 y_5 - x_6 y_6 \exp(\varphi_{R2})}{1 - x_5 y_5 \exp(\varphi_{R2})} \exp(\varphi_{R1})}
\]

Expression above describes the relation of \( E_a \) in term of input signal \( E_1 \). Similarly, expressions for the other components of circulating fields in term of \( E_1 \) can be written as:

\[
E_b = \left[ \exp \left( \frac{\varphi_{R1}}{2} \right) \right] AE_1
\]

\[
E_c = \left[ \frac{j x_3 \sqrt{\kappa_3} \exp(\varphi_{R1})}{1 - x_3 y_3 \exp(\varphi_{R1})} \right] AE_1
\]

\[
E_d = \left[ \frac{j x_5 \sqrt{\kappa_5} \exp(\varphi_{R1})}{1 - x_5 y_5 \exp(\varphi_{R1})} \right] \exp(\varphi_{R2}) AE_1
\]

\[
E_e = \left[ \frac{x_5 y_5 x_6 y_6 \exp(\varphi_{R2})}{1 - x_5 y_5 \exp(\varphi_{R2})} \right] AE_1 \exp(\varphi_{R1})
\]

\[
E_f = \left[ \frac{x_5 y_5 x_6 y_6 \exp(\varphi_{R2})}{1 - x_5 y_5 \exp(\varphi_{R2})} \right] AE_1 \exp(\varphi_{R1})
\]
According to the diagram, circulating field on the center ring E2 can be expressed by:

\[ E_2 = x_3 y_3 E_1 + j x_3 \sqrt{k_3} E_f \tag{16} \]

Substituting Eq. (15) into Eq. (16) yields:

\[ E_2 = \left[ x_3 y_3 - \frac{x_3^2 \exp(\phi_{13})[x_2 y_3 - x_2^2 \exp(\phi_{12})]}{1 - x_2 y_3 \exp(\phi_{13}) - x_2 y_3 \exp(\phi_{14})[x_2 y_3 - x_2^2 \exp(\phi_{15})]} \right] E_1 \tag{17} \]

For simplification, \( C_{12} \) is assigned as:

\[ C_{12} = x_3 y_3 - \frac{x_3^2 \exp(\phi_{13})[x_2 y_3 - x_2^2 \exp(\phi_{12})]}{1 - x_2 y_3 \exp(\phi_{13}) - x_2 y_3 \exp(\phi_{14})[x_2 y_3 - x_2^2 \exp(\phi_{15})]} \tag{18} \]

Finally, expression for \( E_2 \) in term of \( E_1 \) can be written as:

\[ E_2 = C_{12} E_1 \tag{19} \]

Similarly for the left-hand-side part of the circuit as shown in Figure 4, expression for \( E_6 \) in term of \( E_5 \) can be written as:

\[ E_6 = C_{34} E_5 \tag{20} \]

Where

\[ C_{34} = x_4 y_4 - \frac{x_4^2 \exp(\phi_{16})[x_5 y_4 - x_5^2 \exp(\phi_{17})]}{1 - x_5 y_4 \exp(\phi_{16}) - x_5 y_4 \exp(\phi_{18})[x_5 y_4 - x_5^2 \exp(\phi_{19})]} \tag{21} \]

By using these sets of equations, the terms \( E_1, E_3, E_5 \) and \( E_7 \) can be expressed as:

\[ E_1 = \frac{j x_1 \sqrt{k_1} E_{in} \exp(\frac{\phi_f}{4}) + j x_1 y_1 \sqrt{k_3} C_{34} E_{add} \exp(\frac{3 \phi_f}{4})}{1 - C_{12} C_{34} x_1 y_1 x_2 y_2 \exp(\phi_f)} \tag{30} \]

\[ E_3 = \frac{j x_1 \sqrt{k_1} C_{12} E_{in} \exp(\frac{\phi_f}{2}) + j x_1 y_1 \sqrt{k_1} C_{34} E_{add} \exp(\phi_f)}{1 - C_{12} C_{34} x_1 y_1 x_2 y_2 \exp(\phi_f)} \tag{31} \]

\[ E_5 = \frac{j x_1 \sqrt{k_1} C_{34} E_{add} \exp(\frac{\phi_f}{2}) + j x_1 y_1 \sqrt{k_1} C_{12} E_{in} \exp(\phi_f)}{1 - C_{12} C_{34} x_1 y_1 x_2 y_2 \exp(\phi_f)} \tag{32} \]

\[ E_7 = \frac{j x_1 \sqrt{k_1} C_{34} E_{add} \exp(\frac{\phi_f}{4}) + j x_1 y_1 \sqrt{k_1} C_{12} E_{in} \exp(\phi_f)}{1 - C_{12} C_{34} x_1 y_1 x_2 y_2 \exp(\phi_f)} \tag{33} \]

Referring back to Eqs. (1) and (2), by using the sets of equations above, finally expressions for \( E_t \) and \( E_d \) can be written as:

\[ E_t = x_1 y_1 E_{in} + j x_1 \sqrt{k_1} C_{34} E_5 \exp(\frac{\phi_f}{4}) \tag{34} \]

\[ E_d = x_2 y_2 E_{add} + j x_2 \sqrt{k_2} C_{12} E_1 \exp(\frac{\phi_f}{4}) \tag{35} \]

Eqs. (34) and (35) give the expressions for throughput and drop port signals in term of only \( E_{in} \) and \( E_{add} \). The expressions clearly provide direct relation between input and output signals and thus it can be used as the transfer function of double PANDA ring resonator system.

### 4.0 NUMERICAL APPROACH

In this section, transmission properties of optical signals going across nonlinear fiber ring resonator are modelled by using iterative method. Sequences of improving approximate solutions for serially-coupled double ring resonators integrated on both sides of system are generated accordingly. By using Eqs. (3-8) that have been derived earlier, \( E_f \) can be expressed as:

\[ E_f = \left[ \frac{1 - x_3 y_5 \exp(\phi_{R2})}{x_5 y_5 \exp(\phi_{R2})} \right] E_a \tag{36} \]

Eq. (36) describes the relation between circulating field \( E_f \) in term of only \( E_a \). Generally, amplitudes of the electric fields inside the right ring at the \( n \)-th roundtrips are given as:

\[ E_d(n + 1) = x_3 \sqrt{k_3} E_1 \tag{37} \]

\[ E_f(n + 1) = \left[ \frac{1 - x_3 y_5 \exp(\phi_{R2})}{x_5 y_5 \exp(\phi_{R2})} \right] E_a(n + 1) \tag{38} \]

\[ E_d(n + 1) = x_3 y_3 E_1 + x_3 y_3 E_f(n + 1) \tag{39} \]

\[ E_2(n + 1) = x_3 y_3 E_1 + j x_3 \sqrt{k_3} E_f(n + 1) \tag{40} \]
These sets of equations describe the expressions for the circulating and output fields for n number of roundtrips made within the right ring resonator. The term $E_a(1)$ represents the first portion of light coupled into the ring waveguide and it correspond to the first roundtrip of light going across the ring. The field $E_f$ is always one complete roundtrip ahead of $E_a$ as can be seen from the schematic diagram. Due to the symmetrical geometry of PANDA ring resonator system, similar expressions for the left ring can be expressed as:

$$E_a(1) = jx_4\sqrt{\kappa_4}E_5$$  \hspace{1cm} (41) \\
$$E_a(n + 1) = \left[\frac{1 - x_6y_6\exp(\phi_{R6})}{x_6y_6\exp(\phi_{R6})}\right] E_a(n + 1) \hspace{1cm} (42)$$ \\
$$E_a(n + 1) = jx_4\sqrt{\kappa_4}E_5 + x_4y_4E_1(n + 1) \hspace{1cm} (43)$$ \\
$$E_a(n + 1) = x_4y_4E_5 + jx_4\sqrt{\kappa_4}E_1(n + 1) \hspace{1cm} (44)$$

### 5.0 RESULTS AND DISCUSSION

First, basic transmission of double PANDA configuration are simulated and discussed in this section. Output signals ejected from throughput and drop ports are governed by sets of equations that have been first derived in Equations (34) and (35). The mathematical relations suggested that both output power at throughput, $I_t$ and drop port, $I_d$ are extremely dependent on the coupling parameters of first and second coupler on the center ring. Results are generated based on the system with center ring radius of $R=50$ μm and side ring radiuses of $R_1=R_2=R_3=R_4=10$ μm respectively. Coupling coefficients $\kappa_3$, $\kappa_4$, $\kappa_5$ and $\kappa_6$ are fixed at 0.30. Some relations were examined by plotting 3-dimensional graph. During simulation process, input power is set at 1 mW and coupling coefficients values were varied from 0 to 1.

Figure 5 shows the relation between output signals at throughput port and coupling coefficient $\kappa_1$ and $\kappa_2$. From the graph, we can see that $I_t$ reach the highest value at specific point where $\kappa_1$ is equal to one while $\kappa_2$ is zero. Referring to the geometry of double PANDA configuration, as value of $\kappa_1$ increased, more fraction of input light $I_{in}$ are coupled into the ring. This process induces the increasing in the power within the resonator system due to intensity buildup. At the same time, zero $\kappa_2$ means there is no single fraction of light being ejected from the system through the second coupler. Technically, all the circulated fields within the ring will be ejected out through the first coupler and this process directly increased the output power at throughput port which corresponds to $I_t$ with the highest buildup factor recorded at 3.57. The graph also shows that when $\kappa_1$ is kept at zero value, output power of $I_t$ does not changed for any value of $\kappa_2$. This process occurred because there is no fraction of light from input lin being coupled into the resonator system as $\kappa_1=0$, thus the transmission coefficient is now equals to unity and this explains why $I_t$ is having the same value as the initial input.

Figure 6 shows the simulation result of $I_d$ plotted against coupling coefficients $\kappa_1$ and $\kappa_2$ in 3 dimensions. For the output signals at the drop port, maximum power was recorded at the point where $\kappa_1=0$ and $\kappa_2=1$ respectively. Highest buildup factor was calculated at 3.87. Dynamical behaviors of the generated output signals are demonstrated as the peak power of both $I_t$ and $I_d$ varied at different points on the plane as shown in simulation results.

In Figure 7, peak power of the output signal produced at drop-port of double PANDA system is plotted against coupling coefficients $\kappa_5$ and $\kappa_6$. Radiuses of the rings are $R=50$ μm and $R_1=R_2=R_3=R_4=10$ μm. Values of $\kappa_1$ and $\kappa_2$ are fixed at 0.1 and 0.35 while $\kappa_3$ and $\kappa_4$ are kept at constant value of 0.35 respectively. Both variable parameters $\kappa_5$ and $\kappa_6$ are varied from 0 to 1 and the corresponding 3-dimensional result is shown in the Figure 7.
Based on the 3-dimensional result, any increment in the value of $\kappa_5$ and $\kappa_6$ will result in greater peak power of $I_d$ signal. Highest possible output power of this system is achieved when both coupling coefficient values is unity as clearly shown. Thus, it is said that greater values of both $\kappa_5$ and $\kappa_6$ are required to optimize the output signals. In the case where both coupling coefficients are set at zero, there is no component of light from the first ring being coupled into the second ring on either sides of the configuration. Technically, intensity buildup process only occurred in the first ring. In the other words, the outer rings on both sides are neglected in this particular case, which leads to a lower output power recorded at 1.86 mW. This phenomenon shows that the additional rings that are serially coupled on both sides of the system introduce additional intensity buildup on circulated field components. This process results in greater output power as the signal being ejected and collected at throughput and drop port of the system. In this case, mathematical expression that describes the ratio between output and input fields is obtained from Equations (30) to (35):

$$\frac{E_t}{E_i} = \frac{y_1 - y_2 e^{-i\phi_1} y_1 e^{i\phi_1} y_1 e^{i\phi_2} - y_2 e^{i\phi_2}}{1 - y_2 e^{i\phi_2} y_1 e^{i\phi_1} y_1 e^{i\phi_2} - y_2 e^{i\phi_2}}$$

(45)

Buildup factor for this configuration, $B$ is given by squared modulus of Equation (45). This system is assumed to be lossless such that attenuation coefficient is unity for both ring ($a_1=a_2=1$). For simplification purposes, we assumed that both ring are having the same size where $L_1=L_2$ such that $\phi_1=\phi_2=\phi$. To simplify the model further, coupling coefficients are defined as $\kappa_1=\kappa_2=\kappa$. By substituting $t = \sqrt{1-\kappa}$, we can write:

$$B = \frac{2t^2 + 6t^3 + t^4 - 2t^5 \cos \phi + 4t^2 \cos 2\phi - 2t \cos \phi}{1 + 2t^2 + 6t^3 + t^4 - 2t^5 \cos 2\phi - 4t^2 \cos \phi + 2t \cos \phi - 2t^3 \cos \phi}$$

(46)

Considering a system under resonant condition, the terms $\cos(\phi)$ and $\cos(2\phi)$ is equal to unity. Buildup factor can be written in simplest form as:

$$B = \frac{6t^2 + t^4}{(1-t^2)^2}$$

(47)

For the convenience of discussion, we consider a system with coupling coefficient values of $\kappa=0.7$. Calculation shows that the ratio between output and input intensities of this system is equal to 13.82 at resonant. It is clearly shown that buildup factor for this configuration is significantly high due to the buildup of intensity process that occur within the optical resonating system.

Input signal in the form of bright soliton pulse centered at the communication wavelength around 1550 nm and full-width at half maximum (FWHM) around 100 ps is considered in this section. Initial signal's power is 2 mW while radius of the center and sides ring are fixed at 50 μm and 10 μm respectively. The ring waveguide cross sectional area is 0.50 μm². The attenuation loss is 10 dB/mm around 1550 nm and intensity insertion loss γ=0.1 respectively [28]. Effective refractive index $n_{eff}$ is 3.169 while the intensity-dependent index coefficient $n_{2}$ of InGaAsP/InP waveguide is $1.76 \times 10^{-12} \text{cm}^2/\text{W}$ [29]. For further analysis, delay characteristic of output signals ejected from the proposed system are investigated numerically. For comparison purposes, time delay spectrum for light pulse in double PANDA and single ring (SRR) system were calculated and demonstrated as in Figures 8 and 9.
For SRR system, group time delay was recorded around 261 ps with effective bandwidth of 25 pm corresponding to the wavelength of 1550 nm. On the other hand, maximum group time delay for double PANDA system was recorded at 315 ps while bandwidth approaches 47 pm. Large different of group velocity in the entire signal bandwidth induced resonance splitting centered at 1549.97, 1549.98, 1549.99 and 1550 nm that can be observed from the multiple peaks in the light spectrum as clearly shown in Figure 8. This phenomenon arises due to the strong mutual-coupling and time delay contributed from the series of ring resonators coupled on the side of center ring respectively. As demonstrated in Figure 10, group time delay for a pulse propagating in double PANDA system is significantly increased with a noticeable 3-dB bandwidth. This phenomenon can be explained by taking into consideration the dispersive effect induced in optical waveguide that are being coupled to a series of optical resonator. During propagation, group velocity of light pulse is significantly reduced due to its effective circulation in each ring resonator before it pass through the structure. Phase shift acquired by interacting light pulse is greatly dependent on its detuning from cavity resonance. In addition, nonlinear response (Self Phase Modulation) from the interaction of light pulse with nonlinear ring resonator produces additional phase shift. Thus, this structure introduces a large dispersive effect which leads to group time delay and bandwidth modification of light pulse.

Figure 9 Time delay spectrum for light pulse in SRR system

Figure 10 Time delay spectrum for light pulse in double PANDA and SRR systems

Figure 11 shows the temporal profile of signal pulse after experiencing slow-light propagation through SRR and double PANDA system. Signals delay between the two pulses are about 45 ps. Interacting signals experienced buildup of intensity as it effectively circulate within the series of additional rings coupled on the left and right sides of center ring. Buildup factor is recorded around 2.5 for this set-up. Bandwidth modulation of the signal introduces a broader spectral width and simultaneously producing shorter pulse width. As shown in Figure 11, pulse width decreases from 100 ps to 60 ps. However, output signal is assumed to maintain its soliton shape even with the shorter pulse width. The time bandwidth product (TBP) of output pulse is measured around 0.35, which is slightly greater than the value of the transform limit of sech2 shape pulse. Thus, for pulse propagation within double PANDA system, signal delay is achieved with noticeable signals gain and minor distortion as illustrated.

Figure 11 Temporal profile of signal pulses through double PANDA and SRR system
6.0 CONCLUSION

Based on z-transform method, analytical expression of proposed double PANDA ring resonator configuration is derived including its optical transfer function. By numerical simulation, group time delay of signal passing through the system are obtained near communication wavelength of 1550 nm. By comparing this particular system to SRR, slow light time delay for double PANDA system increased to 315 ps with effective bandwidth recorded around 47 pm. Resonance splitting is observed due to the mutual coupling and nonlinear response. Maximum pulse delay time is recorded at 45 ps with noticeable signal gain and buildup factor recorded around 2.5. Output signal is simulated with pulse width around 60 ps with observable minor signal distortion. TBP for this signal is calculated around 0.35. To summarize, the proposed system can be used to obtain a delay pulse with broader bandwidth and relatively high buildup factor. This technique provides a new setup to realizing slow light in ultra-compact photonics circuit.

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